

The Heisenberg–Weil Representation and Fast Wireless Communication

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Joint work with:

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- **Oded Schwartz (EECS, Berkeley)**

(0) Motivation - GPS

- **GPS**



CLIENT WANT: Coordinates of satellite and time delay (enables to calculate distance to a satellite)?

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Fact (GPS)

Client receives

$$R[n] = b \cdot \sum_{k=1}^m \alpha_k \cdot e^{\frac{2\pi i}{N} \omega_k \cdot n} \cdot S[n + \tau_k] + \mathcal{W}[n], \quad n \in \mathbb{Z}_N,$$

$m = \# \text{ paths}$, $\alpha_k \in \mathbb{C}$ *intensity*, $\sum_{k=1}^m |\alpha_k|^2 \leq 1$, $\omega_k \in \mathbb{Z}_N$ *Doppler*,
 $\tau_k \in \mathbb{Z}_N$ *delay, along path k*, $\mathcal{W} \in \mathcal{H}$ *random noise*.

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Problem

Design $S \in \mathcal{H}$, and effective method to extract (b, τ) , $\tau = \min\{\tau_k\}$, using R and S .

Motivation - TIME-FREQUENCY SHIFT

- Simpler scenario

$$R[n] = e^{\frac{2\pi i}{N}\omega_0 \cdot n} \cdot S[n + \tau_0] + \mathcal{W}[n].$$

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Problem (Time-Frequency Shift)

Design $S \in \mathcal{H}$, and method of extracting (τ_0, ω_0) from S and R .

(I) Solution - MATCHED FILTER

Definition

Matched filter

$$\left\{ \begin{array}{l} \mathcal{M}(R, S) : \overbrace{\mathbb{Z}_N \times \mathbb{Z}_N}^{\text{Time-Frequency}} \rightarrow \mathbb{C}, \\ \mathcal{M}(R, S)[\tau, \omega] = \left\langle R[n], e^{\frac{2\pi i}{N} \omega \cdot n} \cdot S[n + \tau] \right\rangle. \end{array} \right.$$

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- Identity

$$\mathcal{M}(R, S)[\tau, \omega] = \mathcal{M}(S, S)[\tau - \tau_0, \omega - \omega_0] + O\left(\frac{NSR}{\sqrt{N}}\right).$$

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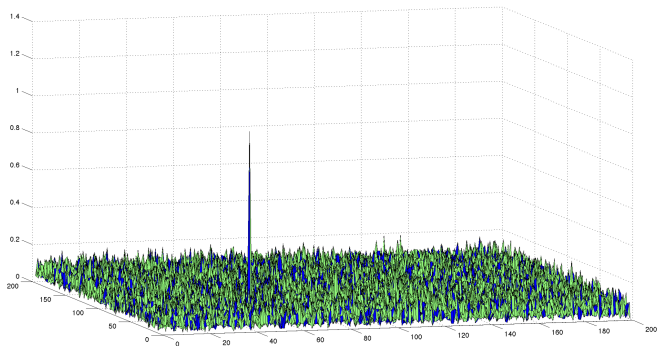
$$\mathcal{M}(R, S)[\tau, \omega] = \mathcal{M}(S, S)[\tau - \tau_0, \omega - \omega_0] + O\left(\frac{NSR}{\sqrt{N}}\right).$$

- Question: What S to use for extracting (τ_0, ω_0) from $\mathcal{M}(R, S)$?

Solution - MATCHED FILTER

- Typical solution: $S = \text{pseudo-random}$.

Example



$|\mathcal{M}(R, S)|$, $S = \text{pseudo-random}$, $(\tau_0, \omega_0) = (50, 50)$.

- Complexity of **Pixel-by-Pixel Algorithm**

$$\underbrace{N}_{\text{for each pixel}} \times \underbrace{N^2}_{\text{\# of pixels}} = N^3.$$

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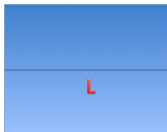
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Problem

Faster algorithm.

MATCHED FILTER - FFT Reduction

- Suppose $L \subset \mathbb{Z}_N \times \mathbb{Z}_N$ line



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$$\mathcal{M}(R, S)[\tau, 0] = \langle R[n], S[n + \tau] \rangle = (S * R)[\tau] \quad - \text{Fast by FFT.}$$

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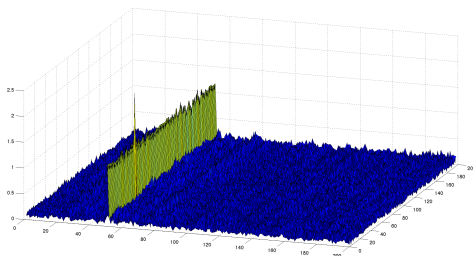
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- Compute entire $\mathcal{M}(R, S)$ in $O(N^2 \cdot \log(N))$ operations.
- **Question:** Can you design S and method to make almost linear number of operations?

(II) Flag Algorithm - IDEA

- Suppose for a line $L \subset \mathbb{Z}_N \times \mathbb{Z}_N$ we construct a signal $S_L \in \mathcal{H}$ with $\mathcal{M}(R, S_L)$ of the form



$$|\mathcal{M}(R, S_L)|, (\tau_0, \omega_0) = (50, 50)$$

Then we have algorithm of complexity $O(N \cdot \log(N))$!!

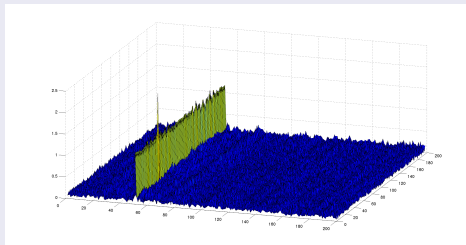
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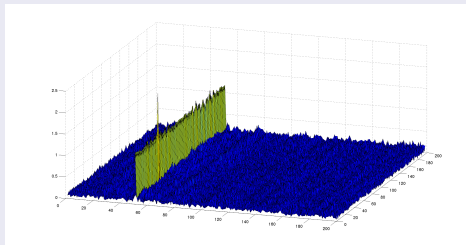


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- 1 **Flag.** Matched filter $\mathcal{M}(R, S_L)$ of the form



$$|\mathcal{M}(R, S_L)|$$

- 2 **Almost orthogonality.** For $L \neq M$ the cross-correlations $|\mathcal{M}(S_L, S_M)[\tau, \omega]| = O\left(\frac{1}{\sqrt{N}}\right)$.

(III) Waveform Design - EXAMPLE

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- $f[n + \tau] = e^{\frac{2\pi i}{N} \tau} \cdot f[n]$, so

$$\begin{aligned} \mathcal{M}(f, f)[\tau, \omega] &= \left\langle f[n], e^{\frac{2\pi i}{N} \omega \cdot n} \cdot f[n + \tau] \right\rangle \\ &= \begin{cases} |\cdot| = 1, & \text{if } \omega = 0; \\ 0, & \text{otherwise.} \end{cases} \end{aligned}$$

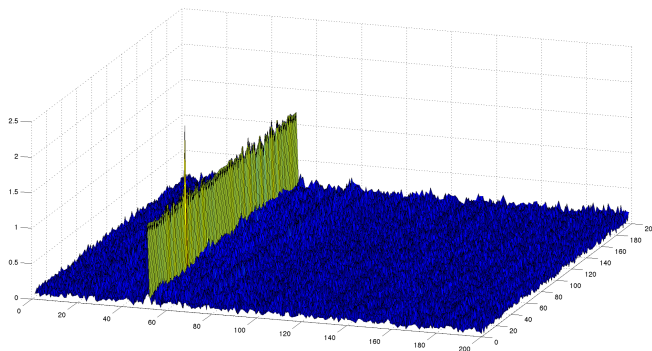
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- Take $S = \underbrace{f}_{\text{exp}} + \underbrace{\varphi}_{\text{pseudo-random}}$, then

$$|\mathcal{M}(R, S)[\tau, \omega]| \approx \begin{cases} 2, & \text{if } (\tau, \omega) = (\tau_0, \omega_0); \\ 1, & \text{on the line } \omega = \omega_0; \\ O\left(\frac{1}{\sqrt{N}}\right), & \text{otherwise.} \end{cases}$$



$$|\mathcal{M}(R, S)|, S = \underbrace{f}_{\text{exp}} + \underbrace{\varphi}_{\text{pseudo-random}}, (\tau_0, \omega_0) = (50, 50)$$

- **Question:** How to generalize the "good" orthonormal basis of \mathcal{H}

$$\mathcal{B} = \left\{ f_a[n] = \frac{1}{\sqrt{N}} e^{\frac{2\pi i}{N} an}; a \in \mathbb{Z}_N \right\} ?$$

Waveform Design - HEISENBERG (LINES) SYSTEM

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- **Answer:** Consider the Heisenberg operators on $\mathcal{H} = \mathbb{C}(\mathbb{Z}_N)$

$$\begin{cases} \pi : \mathbb{Z}_N \times \mathbb{Z}_N \rightarrow U(\mathcal{H}), \\ \pi(\tau, \omega) f[n] = e^{\frac{2\pi i}{N} \omega \cdot n} \cdot f[n + \tau]. \end{cases}$$

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- Property:

$$\pi(\tau, \omega) \circ \pi(\tau', \omega') = e^{\frac{2\pi i}{N} \overbrace{(\tau\omega' - \omega\tau')}^{\det}} \cdot \pi(\tau', \omega') \circ \pi(\tau, \omega).$$

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- Restrict π to the line $L = \{(\tau, 0); \tau \in \mathbb{Z}_N\}$, obtain commutative collection of operators

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Theorem (Linear Algebra – Simultaneous Diagonalization)

There exists a basis for \mathcal{H} of common eigenfunctions

$$\mathcal{B}_L = \{f_\psi; \pi(\tau, 0)[f_\psi] = \overbrace{\psi(\tau)}^{\text{e.v.}} \cdot f_\psi, \text{ for all } \tau \in \mathbb{Z}_N\}.$$

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- Of course $\mathcal{B}_L = \mathcal{B}$.

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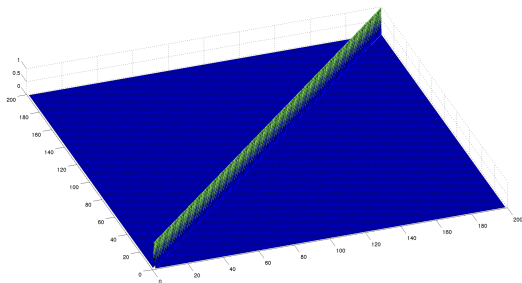
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Theorem (Support, [Calderbank–Howard–Moran, Howe])

We have

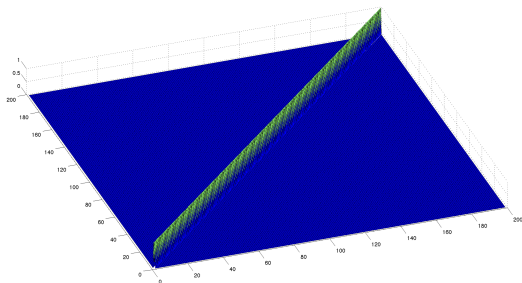
$$|\mathcal{M}(f_\psi, f_\psi)[\tau, \omega]| = \begin{cases} 1, & (\tau, \omega) \in L; \\ 0, & (\tau, \omega) \notin L. \end{cases}$$

HEISENBERG (LINES) SYSTEM – Numerics



• $|\mathcal{M}[f_\psi, f_\psi]|, f_\psi \in \mathcal{B}_L, L = \{(\tau, \tau)\}$

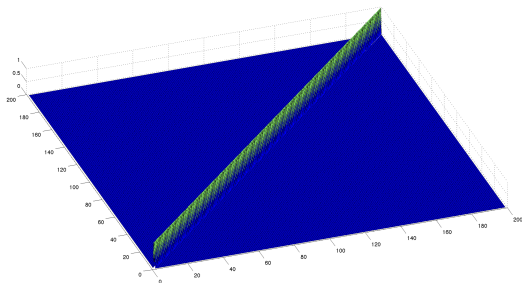
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- What is next?
- Just add any pseudo-random waveform.

Waveform Design - WEIL (PEAK) SYSTEM

- Discrete Fourier transform (DFT) is defined by system

$$\Sigma_W : \quad DFT \circ \pi \begin{pmatrix} \tau \\ \omega \end{pmatrix} = \pi \left(\underbrace{\begin{pmatrix} 0 & -1 \\ 1 & 0 \end{pmatrix}}_W \begin{pmatrix} \tau \\ \omega \end{pmatrix} \right) \circ DFT, \quad \tau, \omega \in \mathbb{Z}_N.$$

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- Special linear group

$$W \in SL_2(\mathbb{Z}_N) = \{g = \begin{pmatrix} a & b \\ c & d \end{pmatrix}; a, b, c, d \in \mathbb{Z}_N, \det(g) = 1\}.$$

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Problem (André Weil)

For each $g \in SL_2(\mathbb{Z}_N)$ find operator $\rho(g)$ on $\mathcal{H} = \mathbb{C}(\mathbb{Z}_N)$, which solves the system of N^2 linear conditions

$$\Sigma_g : \quad \underbrace{\rho(g)}_? \circ \pi \begin{pmatrix} \tau \\ \omega \end{pmatrix} = \pi \left(g \cdot \begin{pmatrix} \tau \\ \omega \end{pmatrix} \right) \circ \underbrace{\rho(g)}_?, \quad \tau, \omega \in \mathbb{Z}_N.$$

Theorem (Stone–von Neumann–Schur–Weil)

$\exists!$ collection $\{\rho(g) \in \text{Sol}(\Sigma_g); g \in SL_2(\mathbb{Z}_N)\}$ such that

$$\begin{cases} \rho : SL_2(\mathbb{Z}_N) \rightarrow U(\mathcal{H}), \\ \rho(g \cdot h) = \rho(g) \circ \rho(h). \end{cases} \quad \text{— Weil representation}$$

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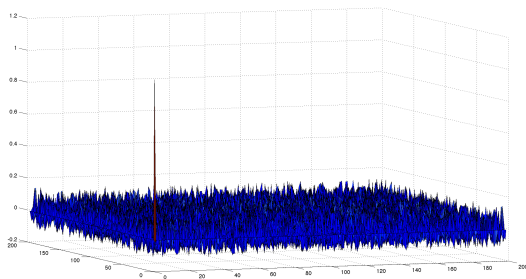
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Waveform Design - WEIL (PEAK) SYSTEM

Theorem (Pseudo-Randomness [G-Hadani-Sochen])

For $\varphi_\chi \in \mathcal{B}_T$ we have $|\mathcal{M}(\varphi_\chi, \varphi_\chi)[\tau, \omega]| = \begin{cases} 1, & (\tau, \omega) = (0, 0); \\ \leq \frac{2}{\sqrt{N}}, & (\tau, \omega) \neq (0, 0). \end{cases}$



$$\mathcal{M}(\varphi_\chi, \varphi_\chi), T = \left\{ \begin{pmatrix} a & 0 \\ 0 & a^{-1} \end{pmatrix} \right\}, \varphi_\chi \in \mathcal{B}_T$$

HEISENBERG+WEIL (FLAG) SYSTEM

Theorem ([Fish–G–Hadani–Sayeed–Schwartz])

Take $S_L = \underbrace{f_L}_{\in \mathcal{B}_L} + \underbrace{\varphi_\chi}_{\in \mathcal{B}_T}$. Then

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① **Flag.** We have $|\mathcal{M}(S_L, S_L)[\tau, \omega]| = \begin{cases} \approx 2, & \text{if } (\tau, \omega) = (0, 0); \\ \approx 1, & \text{if } (\tau, \omega) \in L \setminus (0, 0); \\ \leq \frac{7}{\sqrt{N}}, & \text{otherwise.} \end{cases}$

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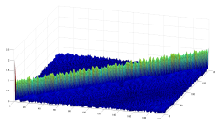
- Flag.** We have $|\mathcal{M}(S_L, S_L)[\tau, \omega]| = \begin{cases} \approx 2, & \text{if } (\tau, \omega) = (0, 0); \\ \approx 1, & \text{if } (\tau, \omega) \in L \setminus (0, 0); \\ \leq \frac{7}{\sqrt{N}}, & \text{otherwise.} \end{cases}$
- Almost orthogonality.** For $L \neq M$ we have $|\mathcal{M}(S_L, S_M)[\tau, \omega]| \leq \frac{7}{\sqrt{N}}$, for every $(\tau, \omega) \in \mathbb{Z}_N \times \mathbb{Z}_N$.

HEISENBERG+WEIL (FLAG) SYSTEM

Theorem ([Fish–G–Hadani–Sayeed–Schwartz])

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$$|\mathcal{M}(S_L, S_L)|, L = \{(\tau, \tau)\}$$

(IV) APPLICATIONS

(A) Channel Estimation

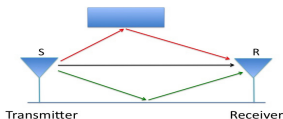
- **Transmit** flag waveform $S = S_L$



(IV) APPLICATIONS

(A) Channel Estimation

- Transmit flag waveform $S = S_L$



- Receive

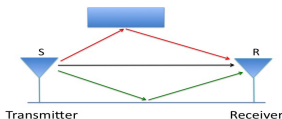
$$R[n] = \sum_{k=1}^m \alpha_k \cdot \underbrace{e^{\frac{2\pi i}{N} \omega_k \cdot n}}_{\text{Doppler}} \cdot S_L[n + \underbrace{\tau_k}_{\text{delay}}] + \mathcal{W}[n],$$

\mathcal{W} = noise, τ_k = delay along path k , ω_k = Doppler along path k ,
 α_k = intensity coefficient, $|\alpha_1|^2 + \dots + |\alpha_m|^2 \leq 1$.

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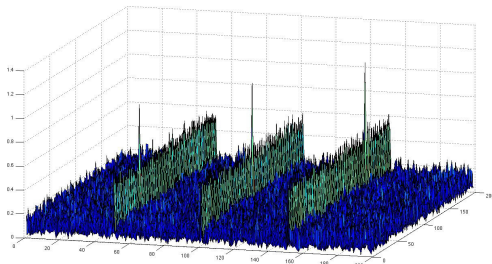
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- **Goal** Extract $(\alpha_k, \tau_k, \omega_k)$'s, using R and S_L .

Application - CHANNEL ESTIMATION

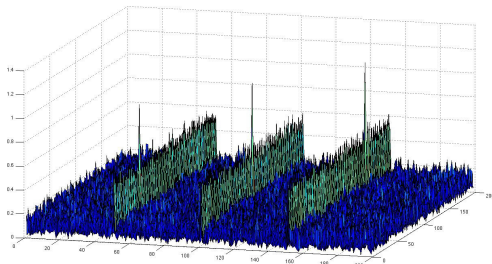
- We have $|\mathcal{M}(R, S_L)[\tau, \omega]| \approx \begin{cases} 2 \cdot \alpha_k, & \text{if } (\tau, \omega) = (\tau_k, \omega_k); \\ 1 \cdot \alpha_k, & \text{if } (\tau, \omega) \in L + (\tau_k, \omega_k) \setminus (\tau_k, \omega_k); \\ \leq O\left(\frac{m}{\sqrt{N}}\right), & \text{otherwise.} \end{cases}$



$$|\mathcal{M}(R, S_L)|, L = \{(\tau, 0)\}, (\alpha_k, \tau_k, \omega_k) = \left(\frac{1}{\sqrt{3}}, 50k, 50k\right), k = 1, 2, 3.$$

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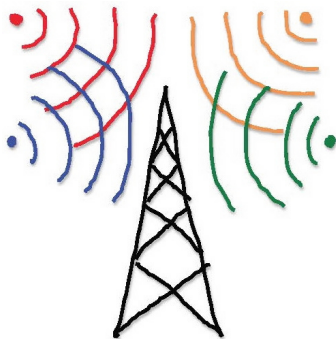
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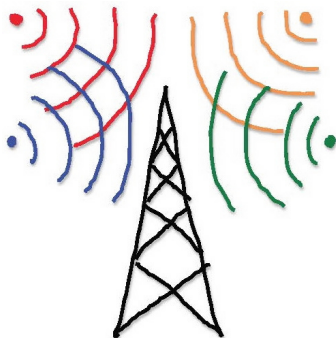
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- Flag method computes channel parameters in $O(m \cdot N \log(N))$.

- (B) Mobile Communication: CDMA, GSM...



- (B) Mobile Communication: CDMA, GSM...



- User transmits message

$$\underbrace{+1 \text{ or } -1}_b \cdot S_L$$

using his private flag waveform S_L .

- Antenna receives:

$$R[n] = b \cdot \sum_{k=1}^m \alpha_k \cdot e^{\frac{2\pi i}{N} \omega_k \cdot n} \cdot S_L[n + \tau_k] + \mathcal{W}[n],$$

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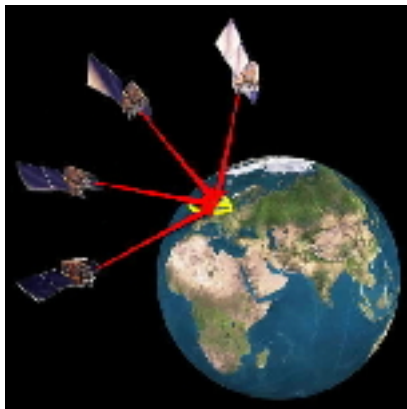
- Antenna receives:

$$R[n] = b \cdot \sum_{k=1}^m \alpha_k \cdot e^{\frac{2\pi i}{N} \omega_k \cdot n} \cdot S_L[n + \tau_k] + \mathcal{W}[n],$$

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- **Goal:** Extract b using R and S_L .
- Method:

$$\left\langle R[n], \sum_{k=1}^m \alpha_k \cdot e^{\frac{2\pi i}{N} \omega_k \cdot n} \cdot S_L[n + \tau_k] \right\rangle \approx \left(\sum_{k=1}^m |\alpha_k|^2 \right) \cdot b.$$

- (C) GPS



CLIENT WANT: Coordinates of satellite and time delay (enables to calculate distance to a satellite)?

Application - GPS

- $S_L, R \in \mathcal{H} = \mathbf{C}(\mathbb{Z}_N), N \gg 1000$.

Application - GPS

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Fact

GPS System

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$\tau_k = \text{delay}$, $\omega_k = \text{Doppler}$, $b = \text{coordinates of satellite}$, $\alpha_k = \text{intensity coefficient}$, $|\alpha_1|^2 + \dots + |\alpha_m|^2 \leq 1$.

Application - GPS

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Problem

Extract (b, τ) , $\tau = \min\{\tau_k \text{'s}\}$ using R and S_L .

Application - GPS

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Problem

Extract (b, τ) , $\tau = \min\{\tau_k \text{'s}\}$ using R and S_L .

Solution

Flag method solves in $O(m \cdot N \log(N))$.

THANK YOU



Sasha



Ronny



Akbar



Oded